

Basel Precision Instruments

Basel Precision Instruments GmbH

Scientific Laboratory Instrumentation Enabling the 2nd Quantum Revolution

Zumbühl lab group meeting
July 16th, 2021

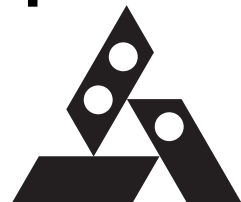
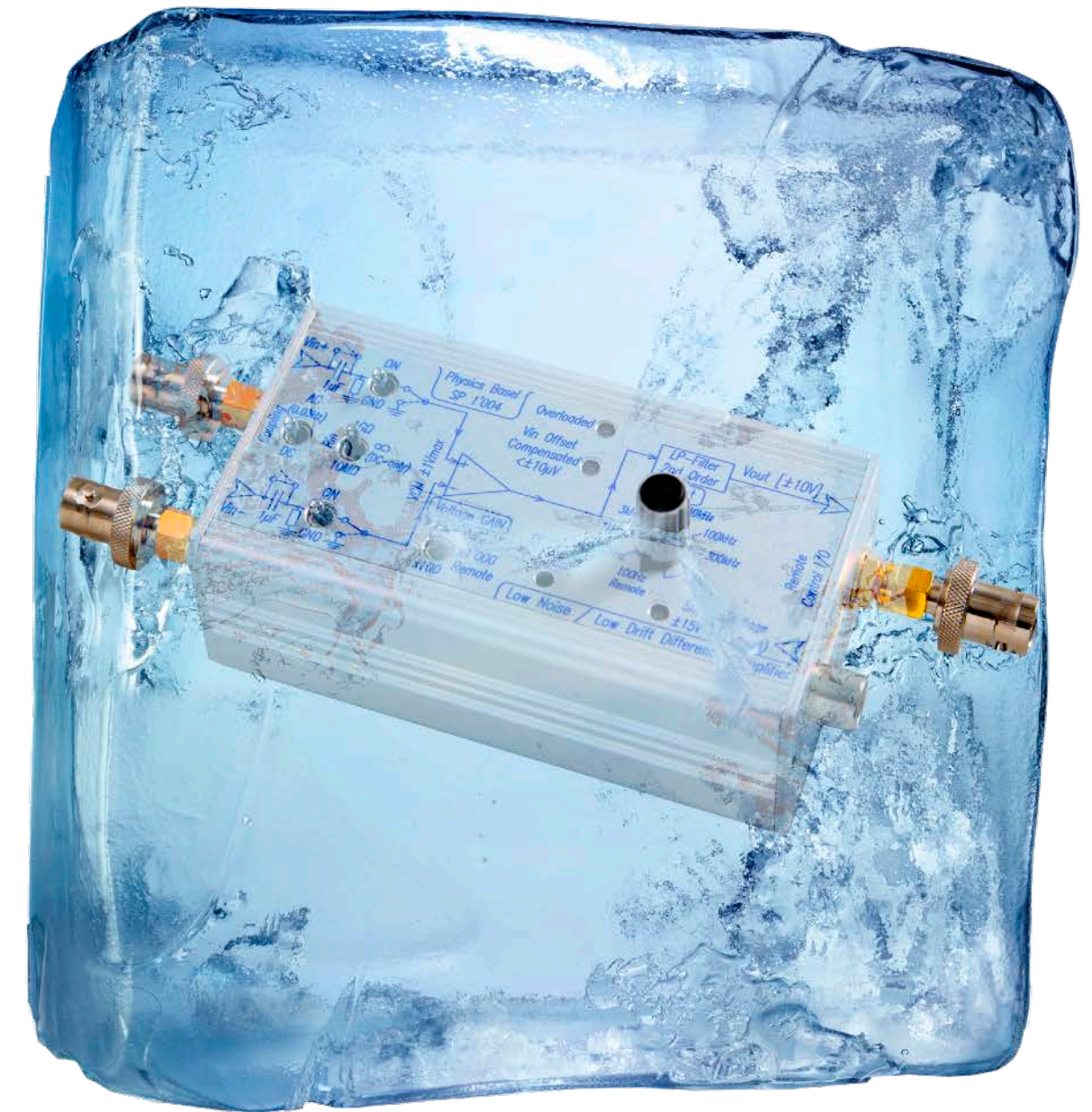
Giving Researchers a Better Lens into the Quantum World

Founded in 2018, BASPI is commercializing the lab instruments behind the **world's coldest chip at < 0.25 mK** at the Zumbühl group at Uni Basel (Samani *et.al.*, APS MM S59.06)

BASPI's premium electronics and cryogenic filters make sensitive measurements possible by:

- minimizing electronic noise
- enhancing stability / minimizing electronic drift and back-action
- lowering sample/electron temperatures

Our instruments make detecting smaller signals & more fragile quantum phenomena possible; hence providing a better lens into the quantum world



World-wide and Expanding Customer Base In the Quantum Market

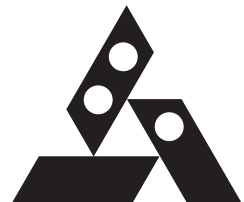
Sold over 100 instruments in 2020 to low temperature research and government laboratories and companies world-wide

Scaling up goal:
4-fold increase in customer
base in the next 2-3 years

Based in Europe - heart of low
temperature research and
technology

Most of our competitors: also
small European SMEs

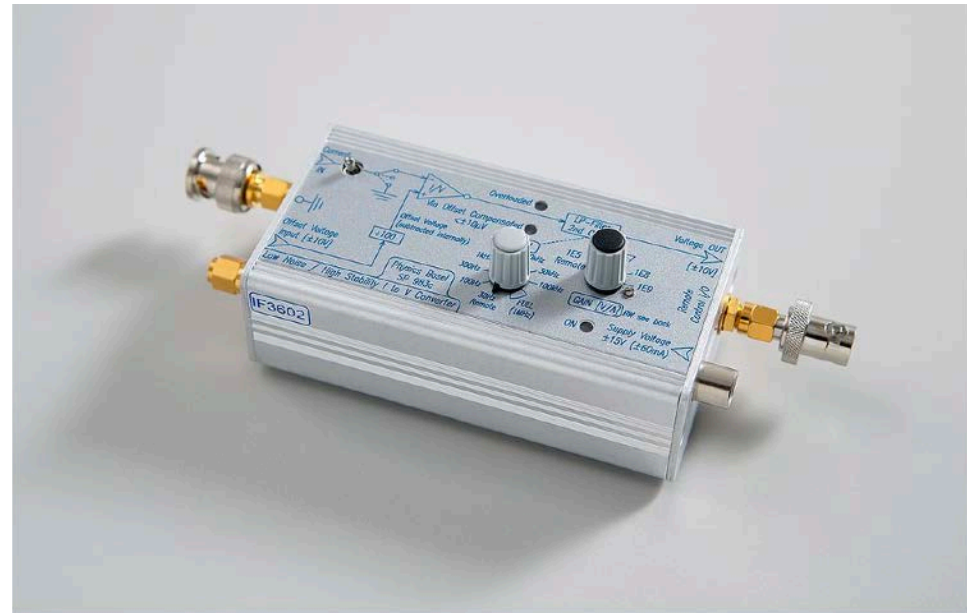
A thorough market research is
underway



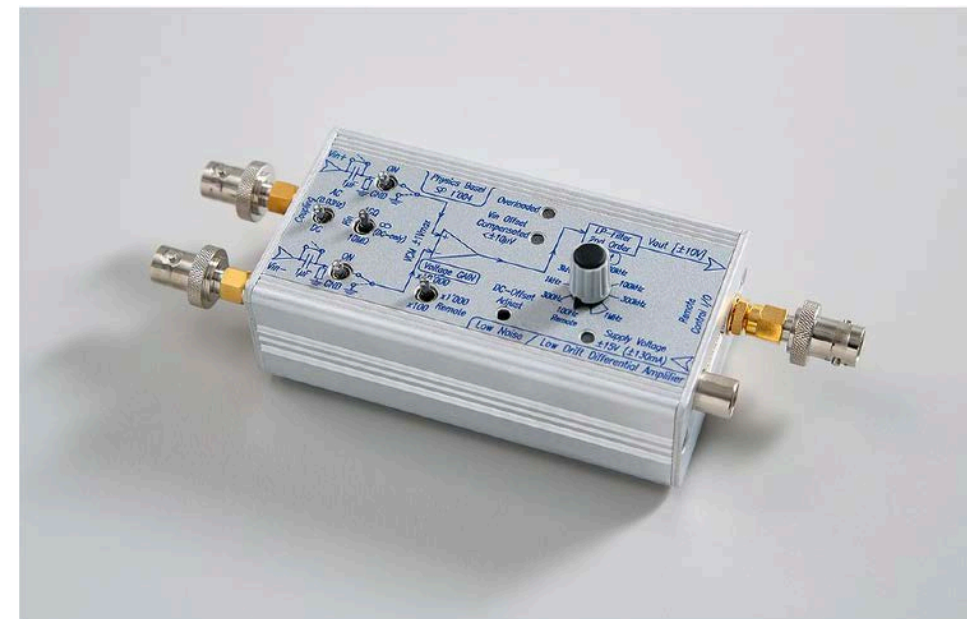
Current Products



Low Noise | High Resolution DAC



Low Noise | High Stability I to V Converter



Low Noise | Low Drift Differential Amplifier



Microwave Filter | Thermalizer

Upcoming Products



Low Noise | High Resolution DAC II



Microwave Filter Thermalizer Tower

BASPI's Compact Ultra-Low Noise Preamplifiers

benefits

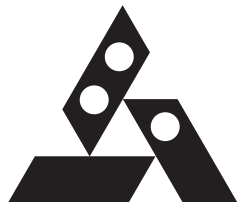
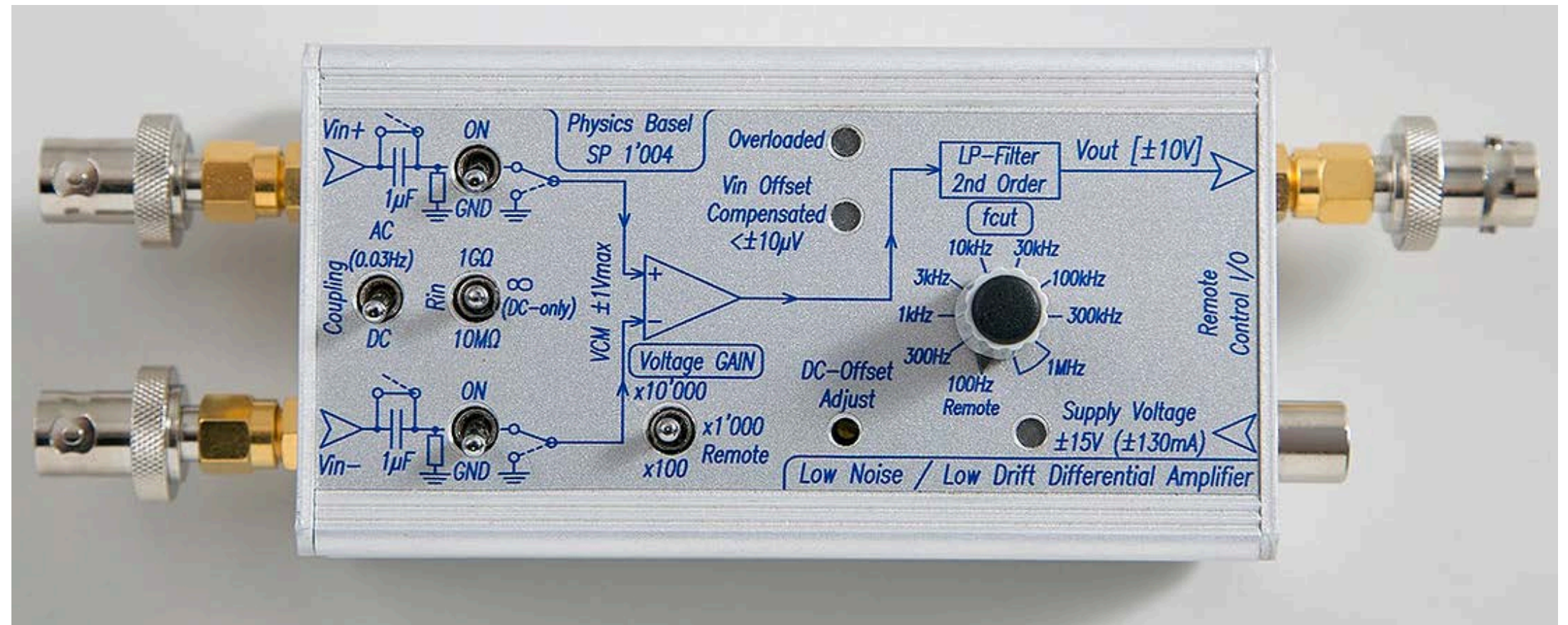
- ultra-low noise preamplification at the first access point
- no additional noise pick up when mounted directly on the breakout box
- minimized back-action on the experiment
- optimized noise performance for specific load
- solves ground loop issues
- possible to apply bias and measure current on the same lead (IV converter)

key features

- compact and low weight
- floating
- ultra-low input voltage and current noise
- several decades of gain
- integrated low-pass-filter
- actively stabilized back action
- input voltage (external) biasing up to 2V (IV converter)



Low-Noise low-drift differential voltage amplifier



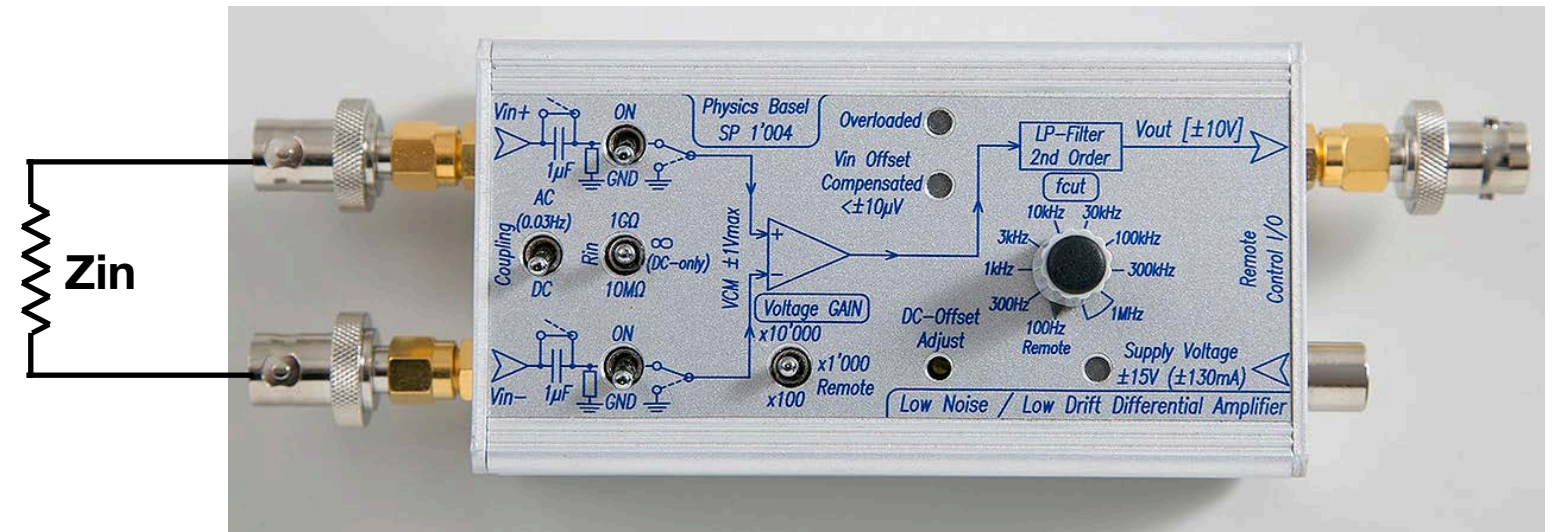
Voltage Preamplifier Noise Sources

Input offset voltage (V_{offset})

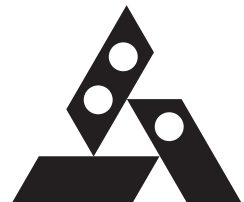
Input voltage noise (U_n)

Input offset or leakage current (I_{offset})

Input current noise (I_n)

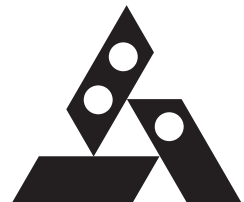
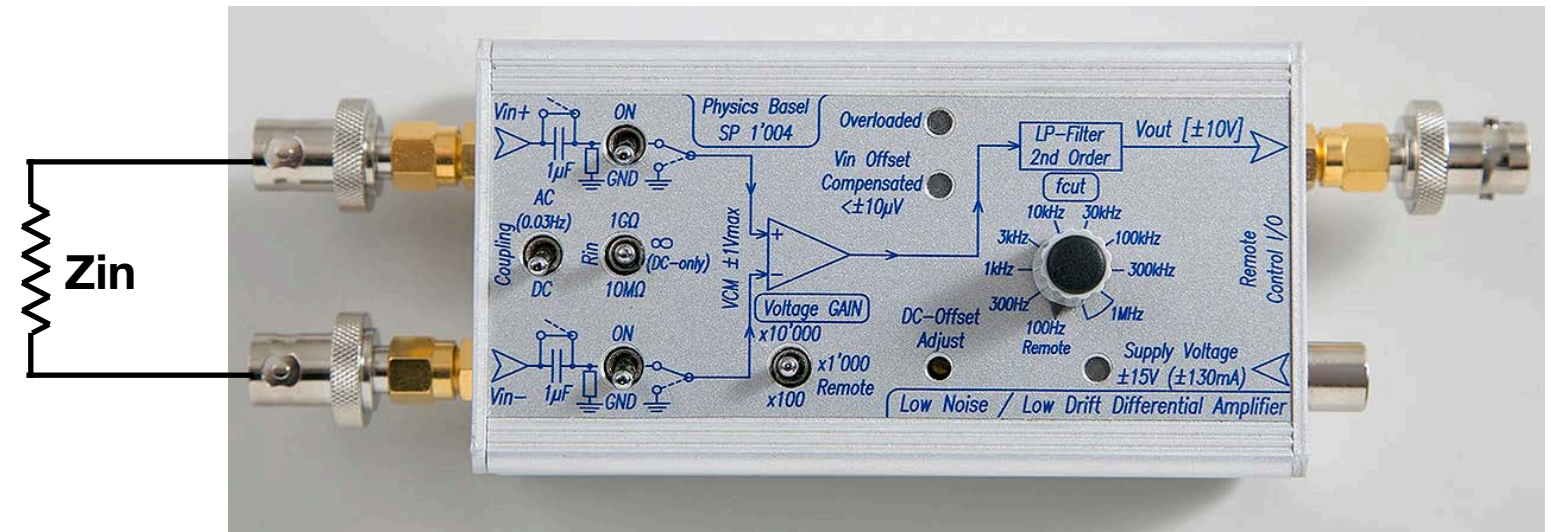


$$V_{\text{output}} = G V_{\text{offset}} + G U_n + G I_{\text{offset}} Z_{\text{in}} + G I_n Z_{\text{in}}$$



Input Offset Voltage (V_{offset})

- Actively compensated and stabilized in BASPi's SP1004 (two decade improvement)
- Can be adjusted manually
- Applied to the sample; relevant for low resistive samples or sensitive bias measurements
- Specs:
 - stabilized voltage within $\pm 10 \mu\text{V}$
 - stabilized drift $< 0.3 \mu\text{V/K}$
 - can be manually adjusted by $\pm 700 \mu\text{V}$



Test Report

Serial Number: SN10040000087

Date: 20/07/2020

Inspector: DM

1. General

Positive Quiescent Supply Current @+15V [mA]:	88	LED ON (green) for Supply Voltage > [V]:	±13.5
Negative Quiescent Supply Current @-15V [mA]:	76	Overload LED (red) @ Vout [V]:	>+9.5 or <-9.5

2. Input Voltage Noise Density (Inputs to GND) | Integrated Noise Voltage

Measuring Frequency [Hz]:	1	10	100	1k	10k	100k
GAIN=100, Input Noise Density [nV/sqrt(Hz)]:	4,65	1,34	1,06	1,05	0,95	0,97
GAIN=1'000, Input Noise Density [nV/sqrt(Hz)]:	4,28	1,29	1,06	0,93	0,94	0,91
GAIN=10'000, Input Noise Density [nV/sqrt(Hz)]:	4,34	1,44	0,94	0,94	0,95	0,87
Integrated Input Noise Voltage 0.5 Hz - 1 kHz (GAIN=1'000):						31,43 nVrms
Integrated Input Noise Voltage 0.5 Hz - 100 kHz (GAIN=1'000):						300,4 nVrms

3. Input Voltage Temperature Drift (Inputs to GND)

Input Voltage Temperature Drift @25°C [µV/K]:	0,345
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4. DC-Offset Adjust Range (with Trimmer)

Input Offset Voltage Range [µV]:	+705,6 to -775,3
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5. DC Input Current

Measured @Input	Vin+	Vin-
Input Current @25°C [pA]:	5,33	5,42

6. Input Current Noise Density, @ Rin = "∞"

Measured @Input	Vin+	Vin-
Current Noise Density @10 Hz [fA/sqrt(Hz)]:	4,54	5,08
Current Noise Density @1 kHz [fA/sqrt(Hz)]:	12,54	14,87

7. Input Diff. Voltage @Offset Comp. ON

Square Input Volt. @0.5 Hz [mVpp]:	115
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8. DC Voltage GAIN Accuracy

Voltage GAIN:	100	1'000	10'000
Mean Error (Offset Corrected) [%]:	1,35	0,07	-0,47

9. Bandwidth and Rise/Fall-Time @fcut = 1 MHz

Voltage GAIN:	100	1'000	10'000
Bandwidth (-3dB, 0.5 Vrms) [kHz]:	1155	1115	1.170
Rise/Fall-Time (10-90%, 1 Vpp) [µs]:	0,296	0,298	0,297

10. Common Mode Voltage

Max. POS [V]:	1,1
Max. NEG [V]:	1,1

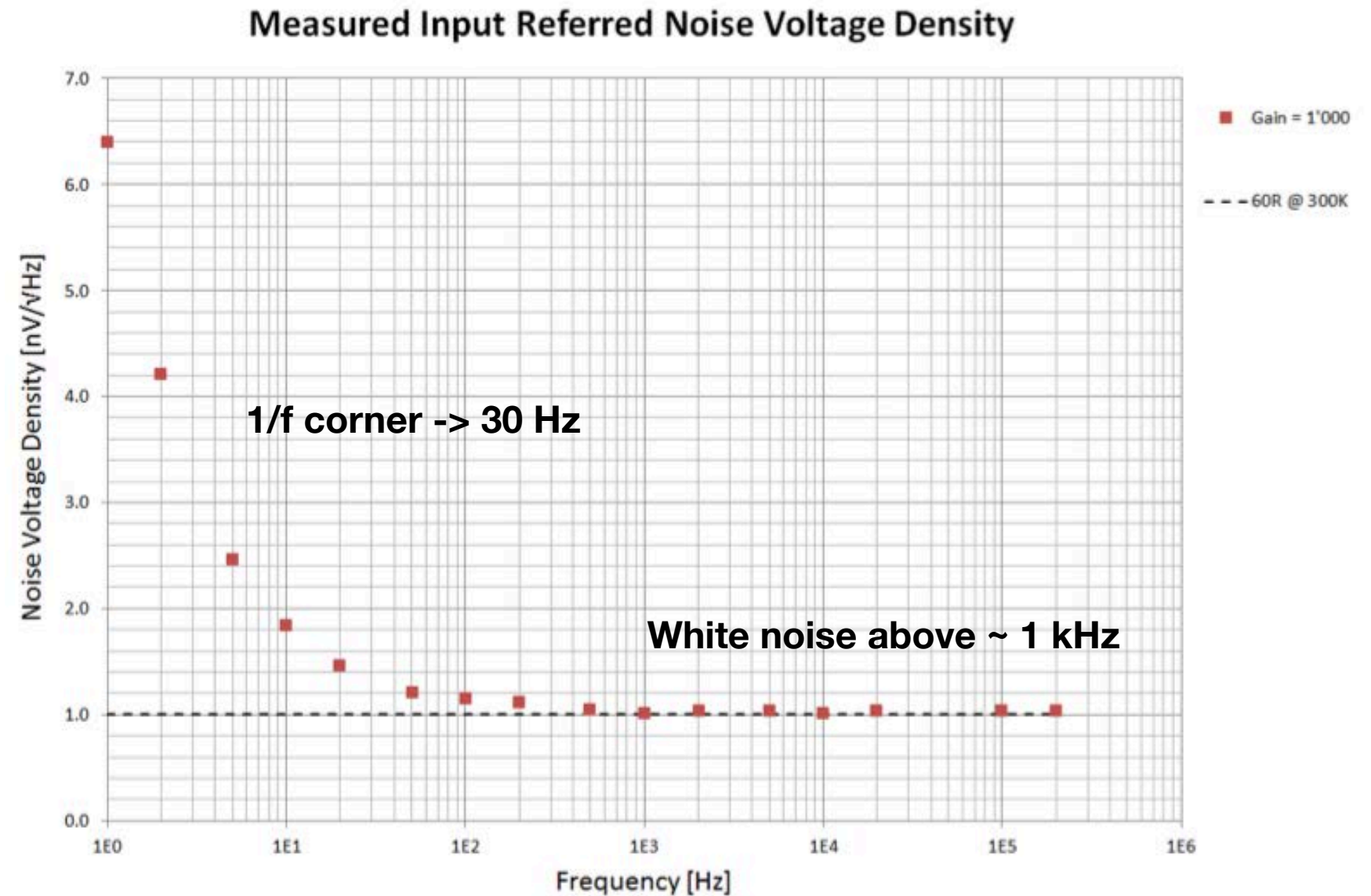
11. Common Mode Rejection-Ratio @GAIN=1'000

Measuring Frequency [Hz]:	0,01	10	100	1k	10k	100k
CMRR @DC-Coupling [dB]:	87,1	136,3	148,5	125,7	104,4	87,5
CMRR @AC-Coupling (Rin = 1 GOhm) [dB]:	NA	113,3	113,6	112,5	102,8	87,4



Input Voltage Noise (U_n)

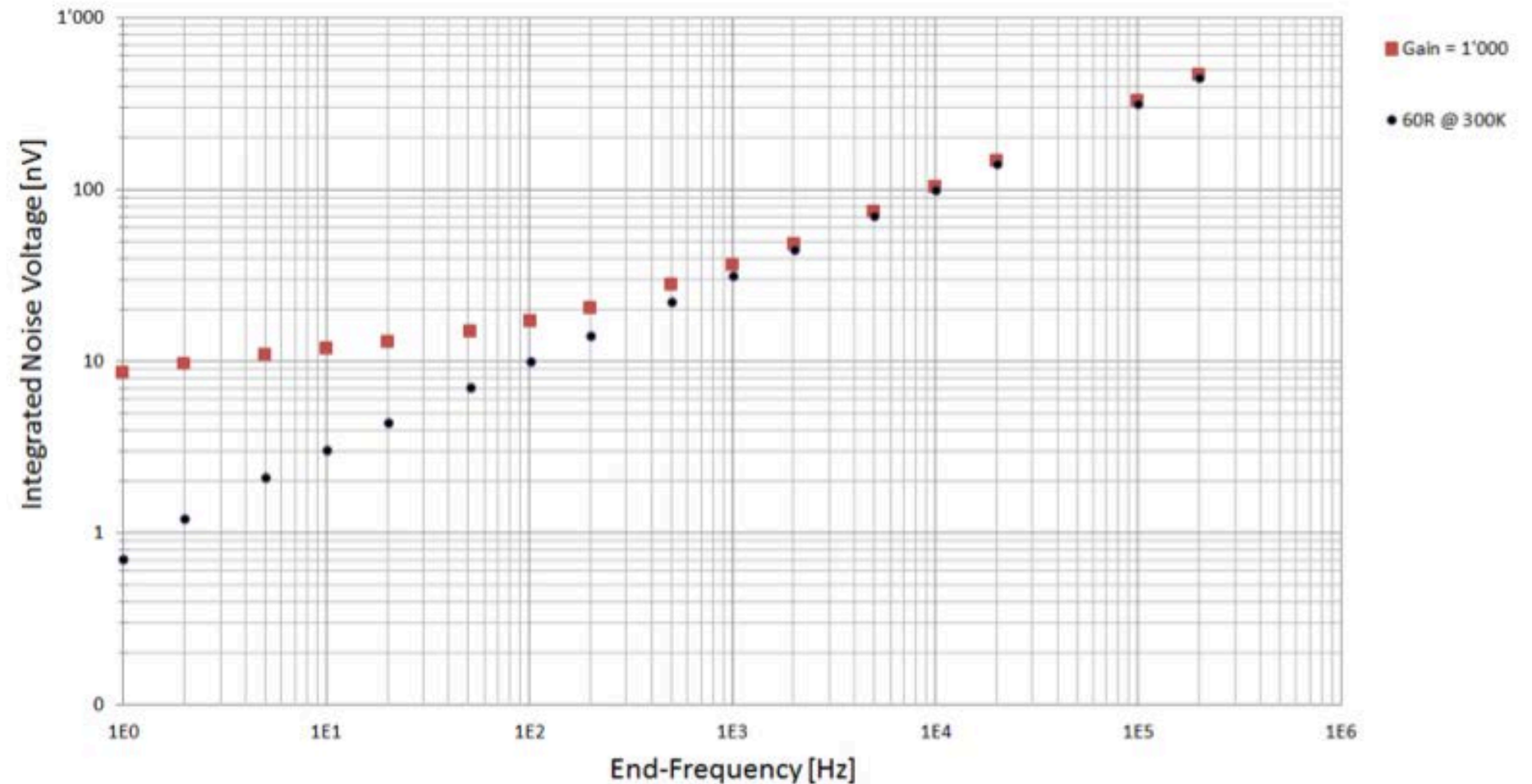
- It is amplified by the gain and contributes to the measurement noise
- Applied to the sample, relevant for low resistive loads
- White noise density of $1 \text{ nV}/\sqrt{\text{Hz}}$ above 1 kHz



Input Voltage Noise (U_n)

- It is amplified by the gain and contributes to the measurement noise
- Applied to the sample, relevant for low resistive loads
- White noise density of $1 \text{ nV}/\sqrt{\text{Hz}}$ above 1 kHz

Measured Integrated Input Referred Noise Voltage (Start-Frequency: 0.5 Hz)



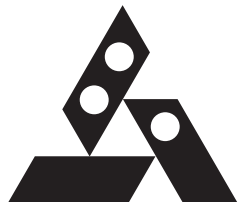
Input Current Noise (In)

- Has two components: shot noise of the leakage current and thermal noise of input resistor
- Can be reduced by selecting low leakage current and by using DC coupling/high input resistor
- Is applied to the sample; relevant for high-resistive samples

$$I_{shot} = \sqrt{2e_0 I_{leak}}$$

$$I_{noise} = \sqrt{\frac{4kT}{R}}$$

Selected Input Resistance [Ohm]	Measured Input Current Noise @ 10 Hz / @30°C [fA/v(Hz)]	Measured Input Current Noise @ 1 kHz / @30°C [fA/v(Hz)]	Theoretical Input Current Noise @ 30°C [fA/v(Hz)]
∞ (DC-only)	6	12	3.1
1 Giga	7	14	5.1
10 Mega	43	48	41.1



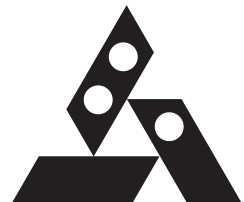
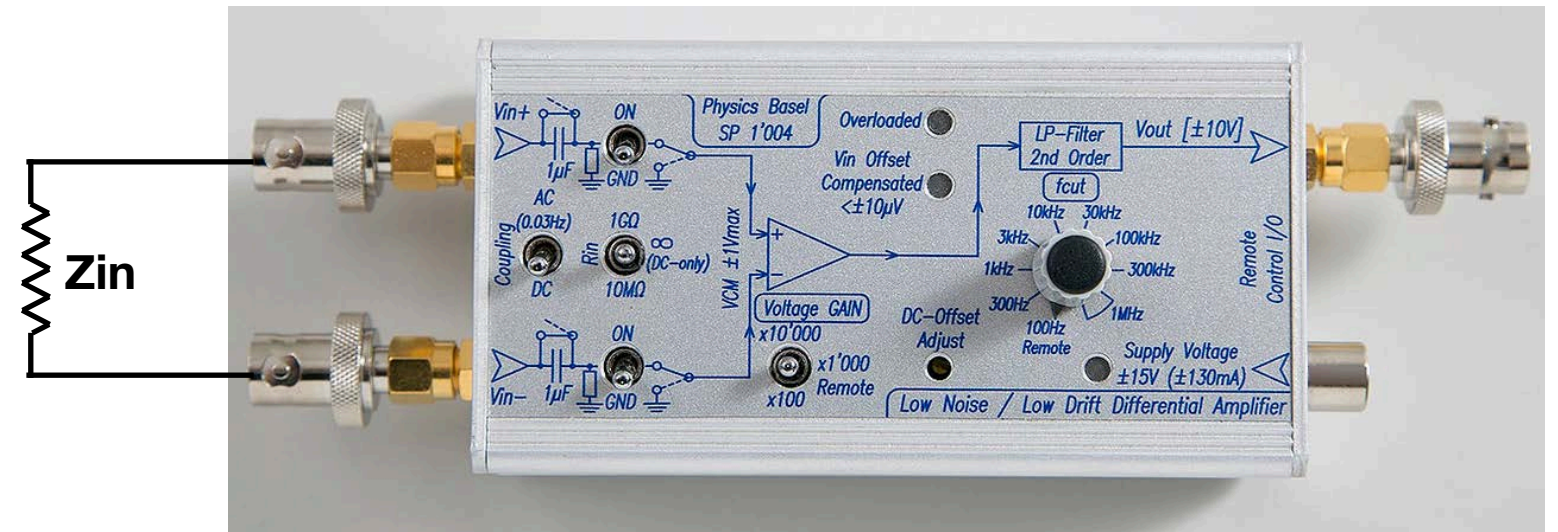
Example:

Measuring the thermal noise of a 1 kΩ resistor at room temperature

- Johnson noise = $4 \text{ nV}/\sqrt{\text{Hz}}$
- Preamp vol. noise = $1 \text{ nV}/\sqrt{\text{Hz}} > 1 \text{ kHz}$
- Select the preamp with lowest voltage noise

Measuring μV signals across a sample with 1 GΩ resistance

- Signal = $\mu\text{V} \cdot \text{G}$
- Current noise = $20 \text{ fA} \cdot 1\text{E}9 \cdot \text{G} = 20 \mu\text{V} \cdot \text{G}$
- Current offset = $15 \text{ pA} \cdot 1\text{E}9 \cdot \text{G} = 15 \text{ mV} \cdot \text{G}$
- Select preamp with lowest current noise, offset and drift



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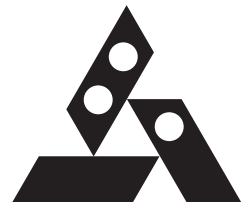
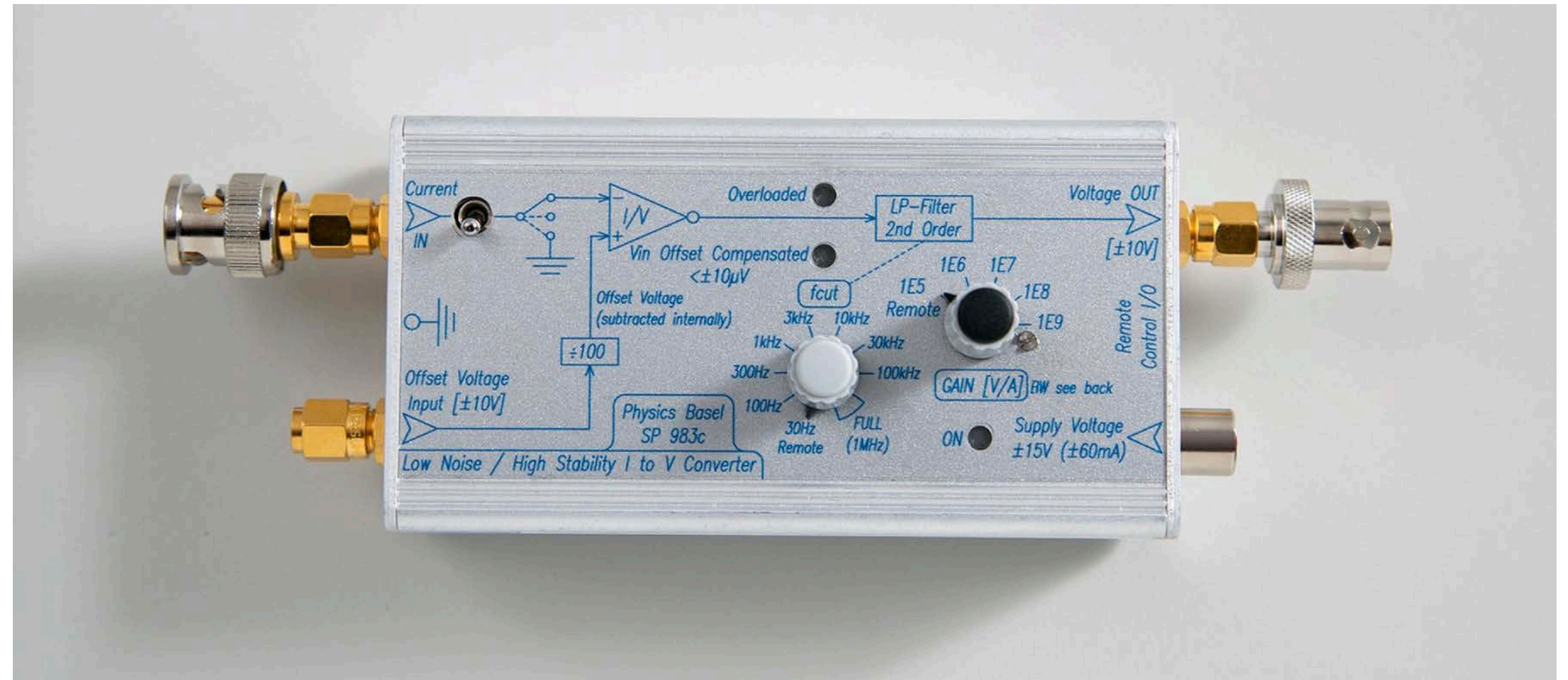
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Rise/Fall-Time (10-90%, 1 Vpp) [µs]:	0,296	0,298	0,297

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Low-Noise high-stability I to V converter



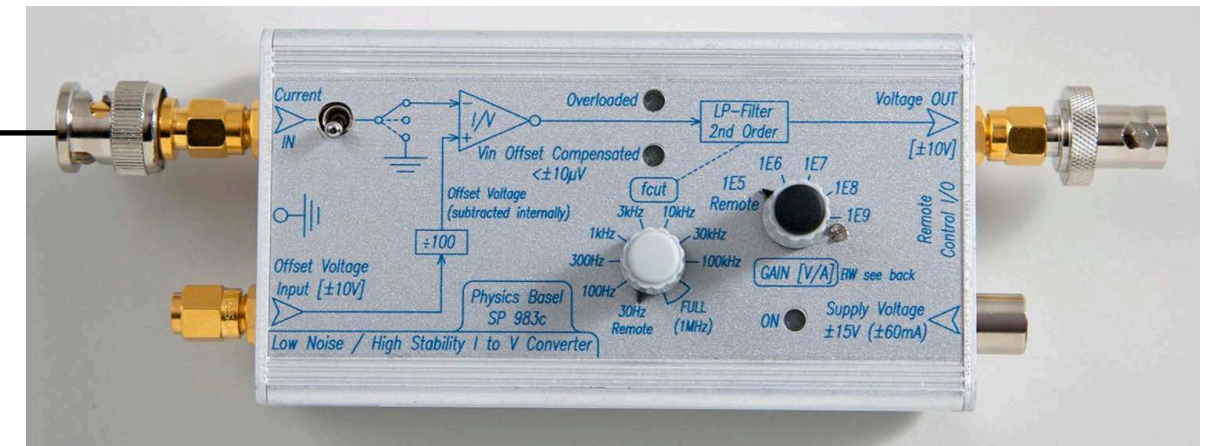
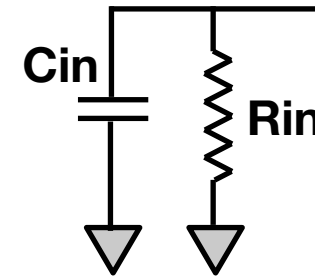
Current Preamplifier Noise Sources

Input offset voltage (V_{offset})

Input voltage noise (U_n)

Input offset current (I_{offset})

Input current noise (I_n)



$$1/Z_{\text{in}} = 1/R_{\text{in}} + j 2\pi f C_{\text{in}}$$

$$V_{\text{output}} = \boxed{\text{Noise-Gain } V_{\text{offset}}} + \boxed{\text{Noise-Gain } U_n} + \boxed{G I_{\text{offset}}} + \boxed{G I_n}$$

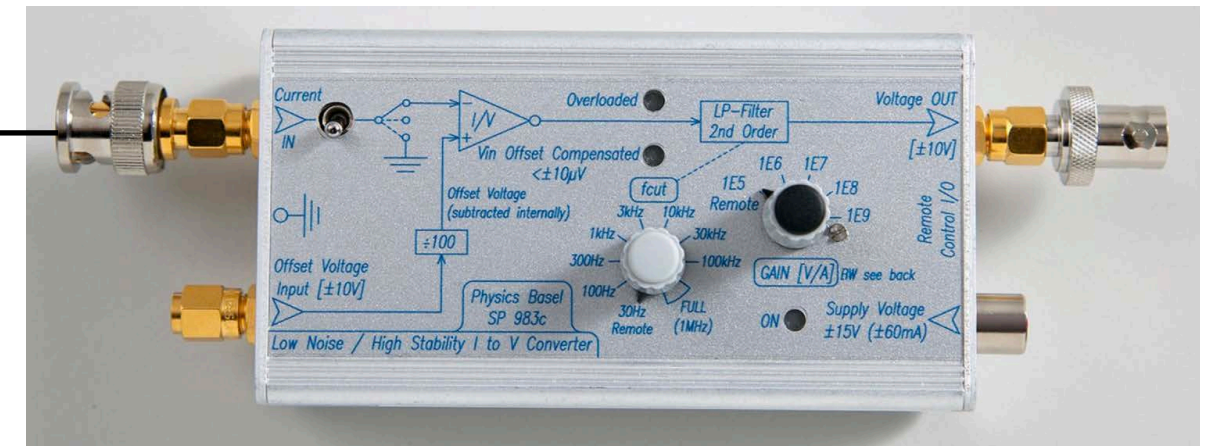
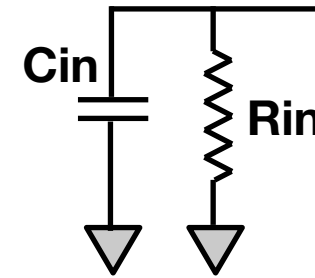
Gain (current gain) = R_f

Noise Gain (voltage gain) = $1 + R_f/Z_{\text{in}} = 1 + R_f/R_{\text{in}} + j 2\pi f R_f C_{\text{in}}$



Input Offset Voltage (V_{offset})

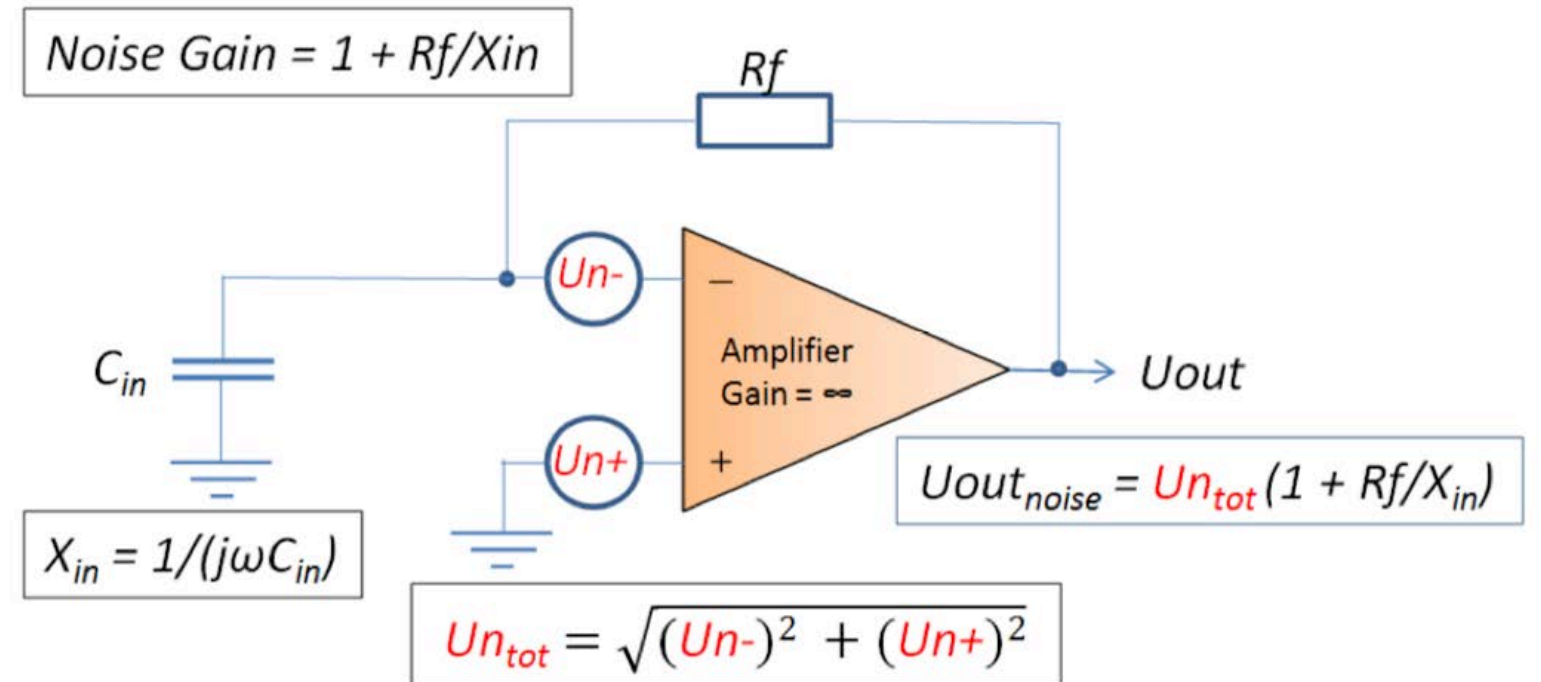
- Actively compensated and stabilized in BASPI's SP983c preamps
- Can be zeroed or shifted using an external bias voltage source
- Is applied to the sample
- Can be used to bias the sample together with an external supply voltage
- Specs:
 - device-specific fixed offset within $\pm 30 \mu\text{V}$
 - stabilized drift $< 0.15 \mu\text{V/K}$ @ 25°C
 - stabilized voltage within $\pm 10 \mu\text{V}$ with respect to fixed or user defined offset
 - Internally subtracted (external bias)



Input Voltage Noise (Vn)

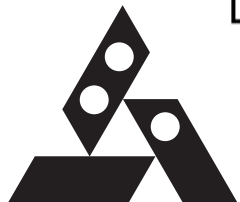
- It is applied to the sample; causes power dissipation, relevant for low-temperature experiments
- Each J-FET has its own voltage noise source (Un_{\pm}) which gets amplified by the noise gain:

Noise Gain = $1 + Rf/Zin = 1 + Rf/Rin + j 2\pi f Rf.Cin$



- Relevant for small load impedance
- Increases linearly with frequency for capacitive loads (also noise peaking)

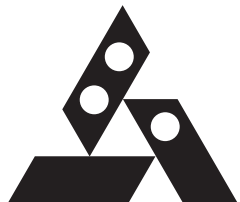
@ f = 10 Hz	@ f = 30 Hz	@ f = 100 Hz	@ f = 1 kHz
2 nV/sqrt(Hz)	1.6 nV/sqrt(Hz)	1.5 nV/sqrt(Hz)	1.2 nV/sqrt(Hz)



Input Current Noise

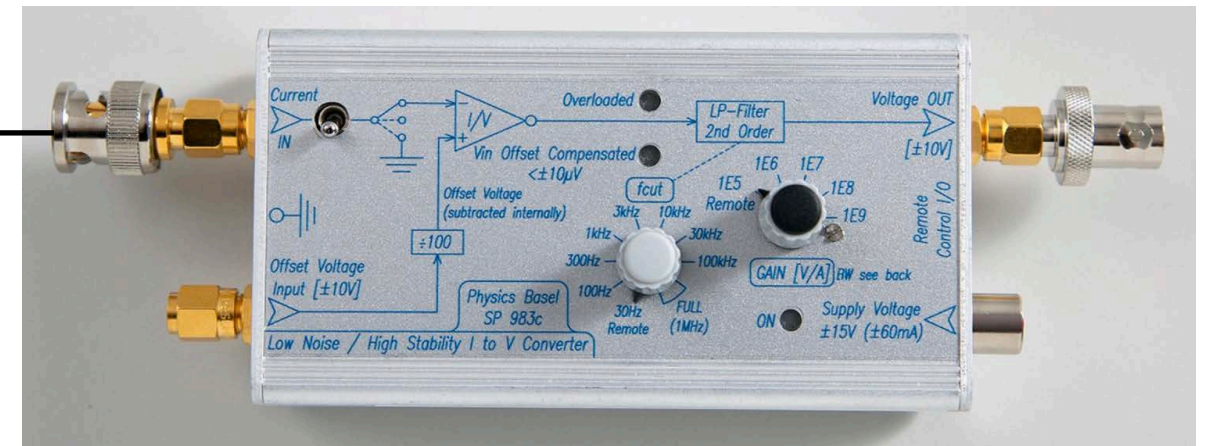
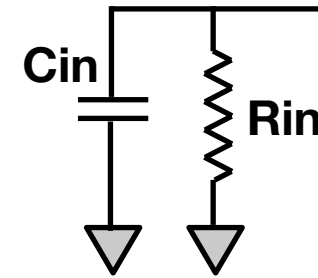
- Originates from the thermal noise of the feedback (Gain) resistor with a $1/\sqrt{R_f}$ dependence
- The signal to noise ratio therefore increases with $\sqrt{R_f}$; hence use the highest gain possible

GAIN [V/A]	Current Noise @ 10 Hz [fA/sqrt(Hz)]	Current Noise @ 1 kHz [fA/sqrt(Hz)]	Theoretical Limit [fA/sqrt(Hz)]
10^9	6	9	4.1
10^8	14	16	13
10^7	42	43	41
10^6	135	140	130
10^5	576	582	410



Input Offset Current

- Originates from the J-FETs
- Does not go through the sample
- Typical value 35 pA



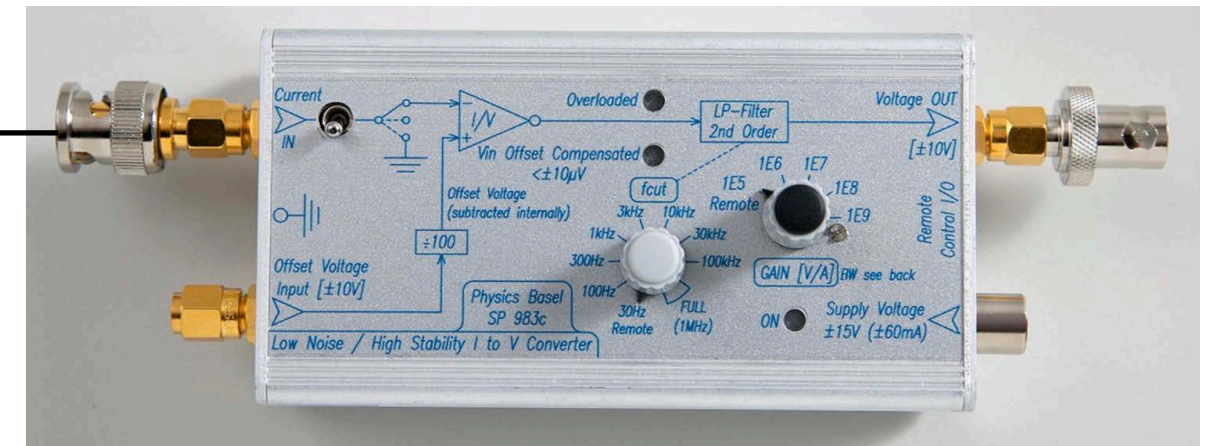
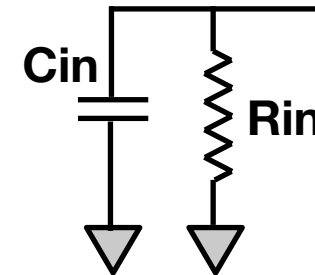
Current Preamplifier Noise Sources

Input offset voltage (V_{offset})

Input voltage noise (U_n)

Input offset current (I_{offset})

Input current noise (I_n)



6 fA/√Hz @10 Hz and $G=10^9$
Leads to 6 μV/√Hz output

$$V_{\text{output}} = \boxed{\text{Noise-Gain } V_{\text{offset}}} + \boxed{\text{Noise-Gain } U_n} + \boxed{G I_{\text{offset}}} + \boxed{G I_n}$$

2 nV/√Hz @10 Hz

Current noise dominates for Noise-Gain < 1000

Gain (current gain) = R_f

Noise Gain (voltage gain) = $1+R_f/Z_{in} = 1 + R_f/R_{in} + j 2\pi f R_f C_{in}$

Voltage noise dominates for $R_{in} < 1 \text{ M}\Omega$ or $C_{in} f > 1 \text{ nF kHz}$



Data Sheet

Model	SP983c	-IF	01-IF	-LSK	02-LSK
Input J-FET		IF3602, best for $R < 1 \text{ M}\Omega$ or $C > 1 \text{ nF}$		LSK389A, best for $R > 1 \text{ M}\Omega$ and $C < 1 \text{ nF}$	
Stable, low-noise and overload protected input current					
Current noise @10 Hz & 10^9 V/A ($\text{fA}/\sqrt{\text{Hz}}$)		6		5	
leakage current magnitude (pA)		40	50 *	3	3 *
Stable, low-drift and low-noise input voltage (low voltage noise relevant for $R < 1 \text{ M}\Omega$)					
Input voltage noise @ 10 Hz ($\text{nV}/\sqrt{\text{Hz}}$)		2.0	2.6 *	4.5	5.0 *
Input voltage noise @ 1 kHz ($\text{nV}/\sqrt{\text{Hz}}$)		1.2	2.0 *	1.9	2.7 *
Input voltage drift		0.15 $\mu\text{V}/\text{K}$ @25°C - feedback stabilized			
Input bias voltage (internally subtracted at output)		$\pm 100 \text{ mV}$	$\pm 1 \text{ V}$ NEW!	$\pm 100 \text{ mV}$	$\pm 2 \text{ V}$ NEW!
Gain		five decades 10^5 to 10^9 V/A - remote controllable			
Integrated low-pass filter		30 Hz to 100 kHz - remote controllable			
Bandwidth		24 kHz @ 10^8 V/A			
DC input impedance		33 Ω – 46 Ω			
GBWP		600 MHz		68 MHz	



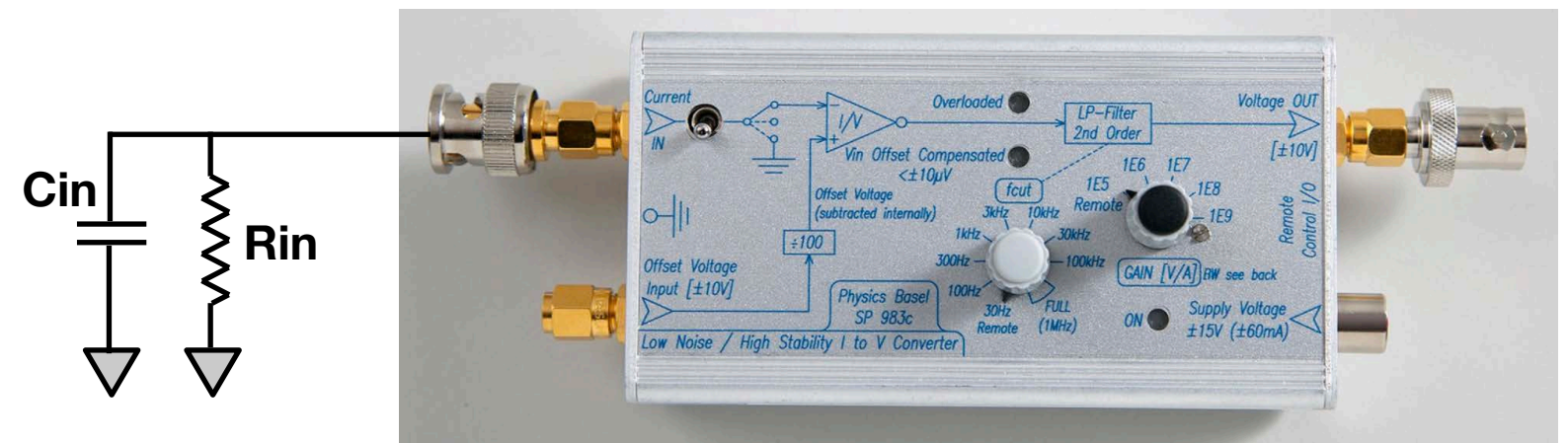
Example:

Measuring pA signal from a highly resistive sample; using Christian's filters on the lines

- $C_{in} = 5 \text{ nF}$, $R_{in} \gg R_f$
- $f = 10 \text{ Hz}$ -> noise-gain = ~ 300
 - Use an LSK with low current noise
- $f = 1 \text{ kHz}$ -> noise-gain = $\sim 30,000$
 - Use IF with low voltage noise

Symmetric Biasing

- How do you select your preamplifiers?



$$\text{Noise Gain (voltage gain)} = 1 + R_f / Z_{in}$$
$$= 1 + R_f / R_{in} + j 2\pi f R_f C_{in}$$



Test Report

Serial Number: SN0983c0000075 with IF602

Date: 13/03/2020

Inspector: D.M.

1. General

Positive Supply Current @+15V [mA]:	41	LED ON (green) for Supply Voltage > [±V]:	13,5
Negative Supply Current @-15V [mA]:	41	Overload LED (red) @ Vout [V]:	9,5

2. Current Input

"Fixed" Input Offset Voltage @25°C [µV]:	8,5	Drift of Input Offset Voltage @25°C [µV/K]:	0,11
DC Input Resistance [Ohm]:	34,7	Gain-Bandwidth Product (GBWP) [MHz]:	583,6
Input Offset Current @25°C [pA]:	49,1	Ext. Offset Voltage Input Step ($\Delta V_{off} < 10 \mu V$) [±V]:	1,3

3. Input Voltage Noise Density

	10	30	100	1.000	Int. Noise Volt [nV _{RMS}]
Measuring Frequency [Hz]:					0.5Hz...1kHz
Input Voltage Noise [nV/sqrt(Hz)]:	1,94	1,59	1,86	1,35	124,2

3. Input Current Noise Density

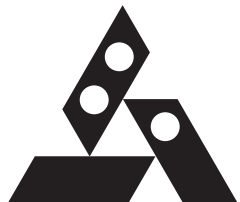
	10E5	10E6	10E7	10E8	10E9
GAIN [V/A]:					
Current Noise @10 Hz [fA/sqrt(Hz)]:	603,6	138,8	45,6	14,6	6,1
Current Noise @1 kHz [fA/sqrt(Hz)]:	593,1	141,2	43,3	15,2	8,4

4. GAIN Accuracy

	10E5	10E6	10E7	10E8	10E9
GAIN [V/A]:					
Accuracy of Gain (Offset Corrected) [%]:	-0,21	0,03	-0,30	-0,25	0,01

5. Bandwidth and Rise/Fall-Time

	10E5	10E6	10E7	10E8	10E9
GAIN [V/A]:					
Bandwidth (-3dB, 1 Vrms) [kHz]:	550	325	101	24,9	1,56
Rise/Fall-Time (10%, 90%) [µs]:	0,623	1,08	3,28	12,9	254,7



Competitive Analysis for LNHS I-to-V Converter

Product name	IVC	PCG-380F	DLPCA-200	Ithaco 1211	SR 570
Manufacturer	BASPI (CH)	Pluto Inst. (KR)	Femto (DE)	DL Instr. (US)	SRS (US)
Variable gain	Up to 10E9 V/A				
Max Bandwidth at 1E9 gain	1.7 kHz	5 kHz	1.1 kHz	4 kHz	15 Hz
Voltage noise density (nV/sqrt Hz) at 1 kHz	1.2	2	4	10	Not specified
Current noise density (fA/sqrt Hz) at 1E9 gain in	5 @10 Hz	4 @10 Hz	4.3 @100 Hz	5 up to 4kHz rms	6 @10 Hz
Input voltage active stabilization	Yes	No	No	No	No
Input voltage drift at room temperature(/degC)	0.15 μ V	1.8 μ V	Not specified	Not specified	Not specified
Possibility to apply bias voltage	Up to 2V	None	Up to 10V	Up to 5V	Up to 5V
Amplifier ground floating	Yes	No	No	No	Yes
Form factor and weight	Compact 165 gr	Semi-compact 350 gr	Compact 320 gr	Bulky 3.7Kg	Bulky 6.7 Kg